# An Introduction to the VNA and Vector Network Analysis

This article provides a brief tutorial on the vector network analyzer, how it works, and its application.

ost design engineers are familiar with tools of the trade such as voltmeters, oscilloscopes, signal generators, and spectrum analyzers. Some may not have had the opportunity to use a vector network analyzer (VNA). It's the intent of this article to introduce VNA measurement and present a few typical applications.

#### Why is a VNA Useful?

Transmission lines can support RF propagation in either direction. Signals traveling along a transmission line may encounter localized impairments that aren't precisely 50  $\Omega$ , such as connectors or transitions from coaxial to planar media (other impedances such as 75  $\Omega$  are common, but for the purposes of this explanation we will use 50  $\Omega$ ). Each impairment generates a reflection that travels back toward the source along the transmission line.

A 50- $\Omega$  load on the end of a 50- $\Omega$ transmission line absorbs all signal energy and reflects nothing. Any load other than 50  $\Omega$  will generate some amount of reflection. The farther the load is from 50  $\Omega$ , the greater the reflection. A short and an open circuit reflect the entire signal back to the source. To characterize reflections, we introduce the *reflection coefficient* ( $\Gamma$ ):

$$\Gamma = \frac{z - z_0}{z + z_0}$$



1. A VNA can separate and measure incident and reflected signals; therefore, it's able to directly determine reflection coefficients. Shown here is a simplified VNA measurement setup.

where  $z_0$  is the characteristic impedance of the source and transmission line and z is the complex impedance of the load.

If z is 0—a short—then  $\Gamma = -1$ , or a complete reflection that's 180 degrees out of phase with the incident signal. If z is infinite—an open—then  $\Gamma = 1$ , or a complete reflection in phase with the incident signal. If  $z = z_0$ —a 50- $\Omega$  load—then  $\Gamma = 0$ , which means no reflection at all.

A VNA can separate and measure incident and reflected signals, and thus directly determine reflection coefficients. *Figure 1* depicts a VNA measurement. The incident signal exits the VNA from Port 1 and arrives at the input to the device under test (DUT). At that interface, there may be some reflection that returns to Port 1. The remainder of the signal passes through the DUT and enters Port 2 of the VNA. Two receivers at Port 1 measure both incident and reflected waves and a receiver at Port 2 measures the wave entering there.

We call the ratio of the incident wave to the reflected wave—the reflection coefficient— $S_{11}$ . The ratio of the signal entering Port 2 to the incident signal leaving Port 1 is called  $S_{21}$ . *Figure 1* shows both.

In the real world, there would be reflections from within the DUT and

another reflection at the output connector, but they're not shown to keep the diagram simple.

There are only a few possibilities for what can happen to the incident signal:

- It can be reflected back to the source in one or more places.
- It can be dissipated as heat within the DUT.
- It can be radiated away by the DUT.
- It can pass through the DUT and make its way to Port 2.

If the signal isn't dissipated or radiated, then a direct relationship exists between  $S_{21}$  and  $S_{11}$ . That is:

$$S_{21}^2 = 1 - S_{11}^2$$

This is simply a conservation of energy statement. Signals that aren't being reflected must pass through the DUT.



 In the actual measurement depicted in Figure 1, a 4-GHz bandpass filter is connected to an SC5090, a 9-GHz two-port analyzer.

## What Does a VNA Measurement Look Like?

In the actual measurement shown below, a 4-GHz bandpass filter is connected to an SC5090, a 9-GHz 2-port analyzer (*Fig. 2*).

The purple trace in *Figure 3* shows  $S_{21}$  in log magnitude format. This is the signal that passes through the filter from Port 1 to Port 2. Markers 2 and 3 indicate the points at which the filter attenuation is 60 dB. Marker 1 is set to the middle of the filter, where most of the signal is passing with little attenuation.

In this chart, the dB scale on the left side is relevant to the red  $S_{11}$  trace, and the zero for  $S_{21}$  is at the top of the screen. The markers show actual values.

The red trace shows  $S_{11}$ , which is the reflection from the filter. In places where the filter isn't passing the signal, it's being completely reflected. In this case, 0 dB is equal to a linear "1," or complete reflection. In the places that have almost no reflection—where  $S_{11}$  is less than -20 dB—the signal is passing through the filter with very little loss.

When speaking of filters, the measurement through the filter—the *insertion loss*—is the  $S_{21}$  reading within the passband and the *return loss* is the  $S_{11}$ measurement in the passband. Here, the insertion loss is 1.2 dB in the middle of the passband, and the return loss is better than 20 dB. Both numbers are traditionally stated as positive values, even though the S-parameters themselves are negative.

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3. Shown are the measurement results for the 4-GHz bandpass filter. *Figure 4* shows measurements being taken on a Wi-Fi antenna. This antenna receives and transmits in the 2.4-GHz band. A one-port measurement of the antenna reveals a sharp dip in  $S_{11}$  at 2.42 GHz and reasonably low values across the measurement range (*Fig. 5*).

Because  $S_{11}$  shows very little reflection—or high return loss—across the Wi-Fi band, the stimulus signal from the VNA must be efficiently radiating away. This antenna is therefore functioning properly.

## The Touchstone Matrix

 $S_{11}$  represents the ratio of the reflection back to Port 1 to the signal emitted by Port 1, and  $S_{21}$  is the ratio of the signal measured at Port 2 to the signal emitted by Port 1. Similarly,



 This is the measurement setup for a Wi-Fi antenna that receives and transmits in the 2.4-GHz band.

 $S_{22}$  is the ratio of the signal reflected back to Port 2 to the signal emitted by Port 2, and  $S_{12}$  is the ratio of the signal measured at Port 1 to the signal emitted by Port 2. These last two might be called *reverse* measurements, as the signal is being emitted by Port 2 instead of Port 1.

If all four S-parameters are known over frequency for a linear two-port DUT, then this represents a complete characterization of the device. The S-parameters, saved in Touchstone format, may be used in a linear simulator to study how the DUT will behave with various RF excitations and loads.

### Conclusion

The VNA measures the RF properties of a DUT by emitting an RF voltage wave and subsequently measures how

> much of it is reflected and how much passes through the DUT for a two-port measurement. For a one-port measurement, it simply measures the reflected wave. A low reflection means that the signal enters the DUT and very little returns.

> Measurements such as that may be performed to determine the function and suitability of an unknown device or to create a Touchstone file for use in linear simulation. This makes the VNA a very powerful tool for the RF engineer.

> Copper Mountain Technologies produces a wide range of VNAs with output frequencies from 1.3 to 330 GHz. Visit <u>www.coppermountaintech.com</u> for product information and educational articles and videos on vector network

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5. A one-port measurement of the antenna in Figure 4 shows a sharp dip in  $S_{11}$  at 2.42 GHz and reasonably low values across the measurement range.



analysis.

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Previously, he was the Manager of RF design at Bird Electronics, where he managed a team of RF Designers and created new and innovative products. Prior to that, he worked for Motorola Component Products Group and was responsible for the design of ceramic comb-line filters for communications devices. Brian graduated from the University of New Mexico, has 40 years of RF Design experience, and has authored three U.S. patents.